## Digital Signal Processor for CD

# HITACHI

ADE-207-162A(Z)

2nd. Edition August 1995

## Description

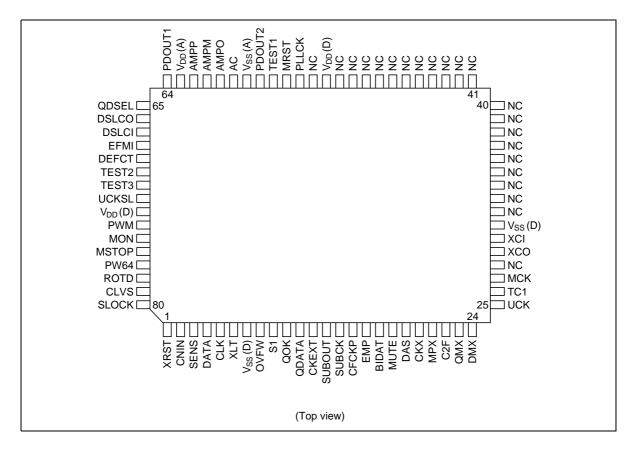
The HD49235FS is a digital signal processor for compact disc (CD) applications.

## Features

- Powerful error correction capability: two-symbol C1 correction and four-symbol C2 correction
- Quadruple-speed reproduction supported (maintaining two-symbol C1 and four-symbol C2 error correction)
- On-chip analog PLL and digital PLL (VCO and phase detector)
- Automatic adjustment of the free-running frequency of the VCO
- Built-in microprocessor interface
- On-chip 80-bit shift registers for Q-code buffering
- Cyclic redundancy check on Q-code values
- Audio output functions: monaural output, single-channel mute, left-right reverse, soft mute, -12-dB attenuation
- 16-kbit RAM on-chip



#### **Pin Arrangement**



## **Pin Description**

Pin						Polarity	
No.	Symbol	Name	I/O*	Connection	Function	Н	L
1	XRST	X (μ-com) reset	I	Microprocessor	Microprocessor interface register reset		Reset
2	CNIN	Counter clock input	I	Servo IC	Pulse input for track counter		
3	SENS	Sensor	то	Microprocessor	Servo status output		
4	DATA	Data	I	Microprocessor	Data input for microprocessor interface		
5	CLK	Clock	l	Microprocessor	Clock input for microprocessor interface		
6	XLT	X (µ-com) latch	I	Microprocessor	Strobe input for microprocessor interface		
7	V <sub>ss</sub> (D)	V <sub>ss</sub> (digital)	—		Digital ground		
8	OVFW	RAM- overflow	0		On-chip RAM overflow signal output		Overflow
9	S1	Subcode sync 1	0	Microprocessor	Subcode sync signal (with protection)		
10	QOK	Q-code OK	0	Microprocessor	Subcode CRC result output	OK	NG
11	QDATA	Q-code data	0	Microprocessor	Subcode Q data output		
12	CKEXT	Clock-EXT	I	Microprocessor	Clock input for Q data readout		
13	SUBOUT	Subcode out	0	CD graphics	Subcode data output for CD graphics		
14	SUBCK	Subcode clock	I	CD graphics	Clock input for SUBOUT subcode readout		
15	CFCKP	C&D frame clock out	0	CD graphics	Subcode frame synchronization signal (7.35 kHz at normal speed, synchronized with PLL)		
16	EMP	Emphasis output	0		Emphasis on/off status output	ON	OFF
17	BIDAT	Biphase date	то		Digital audio interface output		
18	MUTE	Mute	I	Microprocessor	Audio mute input	Mute	
19	DAS	Data serial out	0	DAC or ROM decoder	Serial data output for audio or ROM		
20	СКХ	Clock X	0	DAC or ROM decoder	Strobe clock output for DAS signal		

## **Pin Description** (cont)

	=		
Р	I	n	

Pin						Polarit	y
No.	Symbol	Name	I/O*	Connection	Function	н	L
21	MPX	Multiplex	0	DAC or ROM decoder	Left/right channel switching signal output (44.1 kHz at normal speed, synchronized with DAS)		
22	C2F	C2 flag	0	ROM decoder	C2 error flag output	Error	
23	QMX	Quad multiplex	0		$4 \times MPX$ clock signal (176.4 kHz at normal speed, synchronized with DAS)		
24	DMX	Double multiplex	0		$2 \times MPX$ clock signal (88.2 kHz at normal speed, synchronized with DAS)		
25	UCK	µ-com clock	0	Microprocessor	Clock output for microprocessor (8.5 MHz or 17 MHz)		
26	TC1	Test C1 flag	0		C1 error flag monitor pin		Error
27	MCK	Master clock	0		Master clock output (33.8688 MHz)		
28	NC	No connection	_	Open or VDD	Not connected		
29	XCO	X'tal clock output	ХО	Crystal oscillator	Crystal oscillator output		
30	XCI	X'tal clock input	XI	Crystal oscillator	Crystal oscillator input		
31	V <sub>ss</sub> (D)	V <sub>ss</sub> (digital)	_		Digital ground		
32	NC	No connection	_	Open	Not connected		
33	NC	No connection	_	Open	Not connected		
34	NC	No connection	_	Open	Not connected		
35	NC	No connection	_	Open	Not connected		
36	NC	No connection	_	Open	Not connected		
37	NC	No connection	_	Open	Not connected		
38	NC	No connection	—	Open	Not connected		
39	NC	No connection	_	Open	Not connected		
40	NC	No connection	_	Open	Not connected		
41	NC	No connection	—	Open	Not connected		
42	NC	No connection	_	Open	Not connected		
43	NC	No connection	_	Open	Not connected		
44	NC	No connection	—	Open	Not connected		

## **Pin Description** (cont)

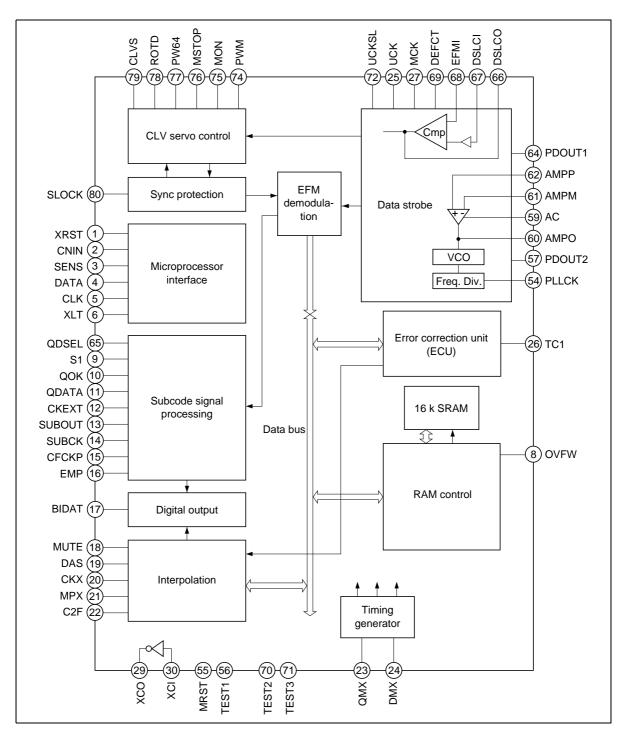
Pin						Polarity	
No.	Symbol	Name	I/O*	Connection	Function	н	L
45	NC	No connection	_	Open	Not connected		
46	NC	No connection	_	Open	Not connected		
47	NC	No connection	—	Open	Not connected		
48	NC	No connection	—	Open	Not connected		
49	NC	No connection	_	Open	Not connected		
50	NC	No connection	_	Open	Not connected		
51	NC	No connection	_	Open	Not connected		
52	V <sub>DD</sub> (D)	$V_{_{DD}}$ (digital)	_		Digital power supply		
53	NC	No connection	_	Open or $V_{\text{\tiny DD}}$	Not connected		
54	PLLCK	PLL clock	0		PLL clock output monitor		
55	MRST	Master reset	IU	Open or $V_{\text{\tiny DD}}$	Master reset of chip		Reset
56	TEST1	TEST 1	IU	Open or $V_{\text{\tiny DD}}$	Test pin		
57	PDOUT2	Phase detect out 2	то	External RC circuit	PLL auto-adjust phase detector output		
58	$V_{ss}$ (A)	V <sub>ss</sub> (analog)	_		Analog ground		
59	AC	Amp compensation	A	External RC circuit	Amplifier phase compensation pin		
60	AMPO	Amp output	AO	External RC circuit	PLL amplifier output		
61	AMPM	Amp minus input	AI	External RC circuit	PLL amplifier inverting input		
62	AMPP	Amp plus input	AI	External RC circuit	PLL amplifier non-inverting input		
63	V <sub>DD</sub> (A)	V <sub>DD</sub> (analog)	_		Analog power supply		
64	PDOUT1	Phase detect out 1	то	External RC circuit	PLL EFM phase detector output		
65	QDSEL	Q-data clock select	IU		Q data readout mode switching signal input	Internal sync	External sync
66	DSLCO	DSL control output	0	External RC circuit	EFM comparator slice level control output		
67	DSLCI	DSL control input	AI	External RC circuit	EFM comparator slice level control input		
68	EFMI	EFM signal input	AI		EFM signal input		
69	DEFCT	Defect	I	Servo IC	Defect detection signal input	Defect	
70	TEST2	TEST 2	IU	Open or $V_{\text{\tiny DD}}$	Test pin		

## Pin Description (cont)

Pin						Polarity	
No.	Symbol	Name	I/O*	Connection	Function	Н	L
71	TEST3	TEST 3	IU	Open or $V_{\text{\tiny DD}}$	Test pin		
72	UCKSL	Microcomputer clock selection	IU		Microprocessor clock switching signal input	16.9344 MHz	8.4672 MHz
73	$V_{_{DD}}$ (D)	$V_{_{DD}}$ (digital)	_		Digital power supply		
74	PWM	Pulse width modulate	то		Constant linear velocity (CLV) control signal for disc motor		
75	MON	Motor on	0		Disc-motor-on status detection output	On	
76	MSTOP	Motor stop	ТО		CLV phase control signal		
77	PW64	Pulse width 64T	0	Microprocessor	Brake release signal		
78	ROTD	Rotate direction	0	Microprocessor	MSB of PWM pin output, for monitoring		
79	CLVS	CLV status	0	Microprocessor	Output indicating normal or starting mode of CLV control	Normal	Starting
80	SLOCK	Sync lock	0	Microprocessor	Disc motor rotation lock signal	Lock	

Note: \* I—input; O—output; IO—input/output; IU—pulled-up input; TO—three-state output; A—analog pin; AI—analog input; AO—analog output; XI—oscillator input; XO—oscillator output

## **Block Diagram**



## **Microprocessor Commands**

Data

										_
Register (Address)	Command	D7	D6	D5	D4	D3	D2	D1	D0	SENS Pin Output
8 (1000)	Mode selections	ROM	ROMEF	DOOFF	SUBCO	SLTSW	0	DCOND	DWIDTH	Z
9 (1001)	Function selections	1	BI1	BIO	WG10TL	SYLCK1	SYLCK0	CRCQ	*	Z
A (1010)	Audio control	MUTEL	MUTER	MONO	ATT	BLGMAIN	BLGSUB	SOFTMT	SWLR	Z
B (1011)	Track counter setting	TC7	TC6	TC5	TC4	TC3	TC2	TC1	TC0	Complete
C (1100)	CLV control	AINTV	ATH	GAIN1	GAIN0	SGAIN1	SGAIN0	PDGAIN1	PDGAIN0	Count
D (1101)	CLV kick control	KICK7	KICK6	KICK5	KICK4	KICK3	KICK2	KICK1	*	Z
E (1110)	CLV mode	ED3	ED2	ED1	ED0	0	*	*	*	BRAKE
F (1111)	ECU mode	0	0	AS0	*	*	*	*	*	Z

Asterisks indicate don't-care bits

#### **Register 8**

		0	1
ROM	D7	Audio (with interpolation)	CD-ROM (no interpolation)
ROMEF	D6	C2 flag output order: lower first	C2 flag output order: upper first
DOOFF	D5	Digital output on	Digital output off
SUBCO	D4	Subcode data not inserted in DAS signal	Subcode data inserted in DAS signal
SLTSW	D3	48-fs clock	64-fs clock
	D2	Normal operation	Illegal setting
DCOND	D1	Condition for switching between digital and analog PLLs: digital PLL when defect detection signal width is 4 frames or more	Condition for switching between digital and analog PLLs: digital PLL when defect detection signal width is 8 frames or more
DWIDTH	D0	Digital PLL termination timing:	
		<ul> <li>8 frames after fall of defect detection signal if width of defect detection signal width is less than 12 frames</li> <li>16 frames after fall of defect detection signal if width of defect detection signal width is 12 frames or more</li> </ul>	<ul> <li>4 frames after fall of defect detection signal if width of defect detection signal width is less than 12 frames</li> <li>8 frames after fall of defect detection signal if width of defect detection signal width is 12 frames or more</li> </ul>

#### **Register 9**

		0	1			
	D7	Illegal setting	Normal operation			
BI1	D6	00: Normal play 0	01: Double-speed play			
BI0	D5	10: Quadruple-speed play	11: Quadruple-speed play			
WG10TL	D4	Sync detection window width:	: ±10 T Sync detection window width: ±19 T			
SYLCK1	D3	Length of time sync lock state	e is maintained when sync signal is missing			
SYLCK0	D2	00: 2 frames 01: 4 frame	9S			
		10: 8 frames 11: 12 fram	les			
CRCQ	D1	QOK flag is not inserted in Q	DATA output QOK flag is inserted in QDATA output			

#### **Register** A

		0	1
MUTEL	D7	Left-channel mute off	Left-channel mute on
MUTER	D6	Right-channel mute off	Right-channel mute on
MONO	D5	Stereo	Monaural
ATT	D4	Attenuation off	Attenuation (-12 dB) on
BLGMAIN	D3	00: Stereo	01: Bilingual, right channel
BLGSUB	D2	10: Bilingual, left channel	11: Bilingual, left channel
SOFTMT	D1	Soft mute off	Soft mute on
SWLR	D0	Normal	Left-right reverse
Notes: 1	Priority	for mute and attenation as fo	llows

Notes: 1. Priority for mute and attenation as follows. "Mute" port > SOFTMT > MUTE L, MUTE R > ATT

In the case of setting "ROM" = 1 (CD-ROM mode), the data of register "A" is ignored and is considered all zero.
 It is recovered as it were, after setting "ROM" = 0.

3. "BLGMAIN" and "BLG SUB" commands are ignored if "SWLR" = 1, and set stereo.

#### **Register B**

	D7	D6	D5	D4	D3	D2	D1	D0
Track counter setting	TC7	TC6	TC5	TC4	TC3	TC2	TC1	TC0
	128	64	32	16	8	4	2	1

#### **Register** C

		0		1
AINTV	D7	Sync detection intervals	on count is tested at 32-frame	Sync detection count is tested at 64-frame intervals
ATH	D6	Sync must be	e detected 4 times or more	Sync must be detected 8 times or more
GAIN1	D5	Speed error	(PWM pin output) gain in CLV	steady state operation
GAIN0	D4	00: –6 dB	01: 0 dB	
		10: +6 dB	11: 0 dB	
SGAIN1	D3	Speed error	gain and access	
SGAIN0	D2	00: –6 dB	01: 0 dB	
		10: +6 dB	11: 0 dB	
PDGAIN1	D1	CLV phase e	error (MSTOP pin output) gain	
PDGAIN0	D0	00: –6 dB	01: 0 dB	
		10: +6 dB	11: 0 dB	

#### **Register D**

	D7	D6	D5	D4	D3	D2	D1	D0
CLV kick control	KICK7	KICK6	KICK5	KICK4	KICK3	KICK2	KICK1	*
(PWM duty cycle)	64/128	32/128	16/128	8/128	4/128	2/128	1/128	*

Asterisks indicate don't-care bits

#### CLV Mode (Register E)

ED3 to ED0/HEX Mode		Mode	Status	
0000	0	STOP	Motor stop	
0110	6	PLAY	Starting mode	
			Normal mode	
1000	8	ROT	Disc motor driven forward	
1001	9	KICK	Kick control	
1010	А	BRAKE	Disc motor driven in reverse	
1100	С	ACS	Access mode	
1110	E	START	Forced starting mode	
1111	F	NORM	Forced normal mode	

#### ECU Mode (Register F)

AS0	Mode	Status
0	FULL	Error correction: C1—two symbols; C2—four symbols
1	E4IHD	C2—four symbol error correction inhibited on track jump

## **Functional Description**

#### Data Strobe

The main functions of this block are described below.

1. Generation of Basic Crystal Clock

**XCI:** Is the inverter input pin for the crystal oscillator.

**XCO:** Is the inverter output pin for the crystal oscillator.

A 33.8688-MHz crystal oscillator clock signal is generated at the XCI and XCO pins. Figure 1 shows the standard external components when a 33.8688-MHz crystal is used.

2. Generation of Basic PLL Clock

**PLLCK:** This is an output pin used for monitoring the VCO oscillator signal. When the PLL is in lock, the frequency is 4.3218 MHz at standard speed, 8.6436 MHz at double speed, or 17.2872 MHz at quadruple speed.

**PDOUT1:** This is a phase detector output pin, for use in data strobing. This pin is in the highimpedance state in the CLV stop mode. In other CLV modes, this pin outputs the result of phase detection in a phase-locked loop formed with the VCO and the EFM signal input at the EFMI pin.

**PDOUT2:** This is a phase detector output pin, for use in adjusting the free-running frequency of the VCO. In CLV stop mode, this pin outputs a pulse-width modulated waveform equivalent to the phase error in a phase-locked loop formed with the VCO and a crystal-oscillator-derived clock signal. In other CLV modes, this pin maintains a pulse-width modulated output with the same duty cycle as in stop mode.

AC: Connect a capacitor for phase compensation of the amplifier.

**AMPO:** Amplifier output pin.

AMPM: Inverting input to the amplifier.

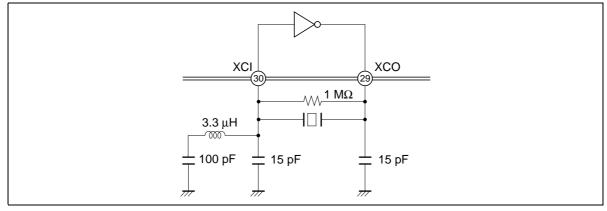


Figure 1 33.8688-MHz Crystal Oscillator Circuit

AMPP: Non-inverting input to the amplifier.

This chip uses a PLL for recovery of the bit clock. A built-in circuit automatically adjusts the freerunning frequency of the PLL, so fewer adjustments are required on the production line. The chip can be forced to adjust its own free-running frequency whenever power is turned on or the speed is changed by switching to CLV stop mode. Thus the free-running frequency is always set to the center of the lock frequency range even if changes occur in the VCO and external circuit constants due to aging.

The principle and usage of automatic adjustment of the free-running frequency will be described below.

- a. In automatic adjustment of the VCO free-running frequency, this chip uses the disc stop signal. The disc stop signal is turned on when the microprocessor writes 0000 in bits ED3, ED2, ED1, and ED0 of register E in the chip's microprocessor interface. (See section 6, Microprocessor Interface.)
- b. When the disc stop signal is turned on, counter (A) in figure 2 becomes a divide-by-98 counter, switch (A) is connected to the output from the VCO, and switch (B) is connected to digital 0. At this time, the circuit for the PDOUT1 output is stopped, so the output of the LPF1 connected to PDOUT1 goes to the fixed DC bias level, which is  $1/2 V_{DD}$ .

The loop formed by PDOUT2  $\rightarrow$  LPF2  $\rightarrow$  amplifier  $\rightarrow$  VCO  $\rightarrow$  counter (A) now operates to lock the VCO oscillator frequency to 34.5744 MHz, which is 8 times the standard CD bit rate (4.3218 MHz).

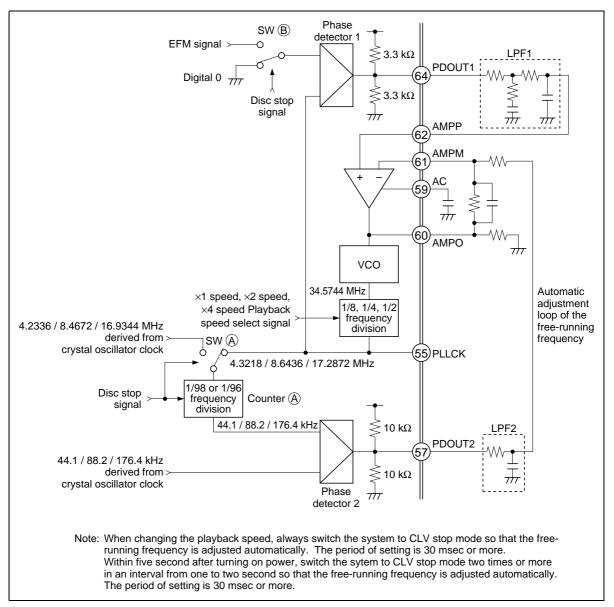


Figure 2

c. Next, reproduction from the disc will begin. When the microprocessor sends command data to start disc rotation, the disc stop signal is turned off.

Switch (A) is now connected to receive a 4.2336 MHz clock from the crystal oscillator and counter (A) is changed to be a divide-by-96 counter. The counter (A) output is held at 44.1 kHz by switching the divisor. Here, switching is performed on the counter (A) output edge. The phase error existing between the VCO and the crystal oscillator clock (the phase error that was detected at adjustment step 2) is maintained.

d. When the disc is rotating and reproduction starts, switch (B) is connected to receive the EFM signal. The phase detector PDOUT1 in figure 2 compares the phases of the EFM signal, which was converted to binary by the EFM comparator, and the bit clock and outputs phase comparison information from the PDOUT1 pin.

As shown in figure 3, when the EFM signal is in phase with the clock produced by the VCO, the PDOUT1 output is high and low for equal lengths of time. When the EFM signal leads the on-chip VCO clock, the high length is longer than the low length. When the EFM signal lags the on-chip VCO clock, the high length is shorter.

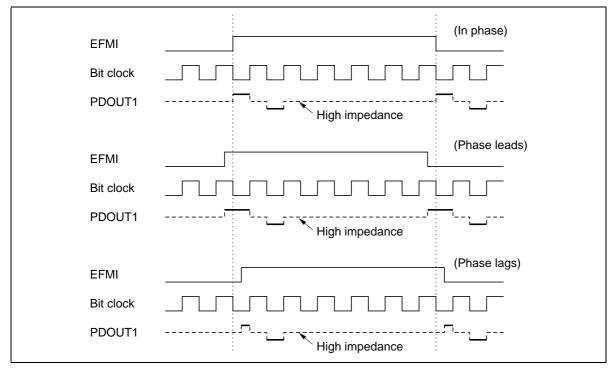


Figure 3 Timing of PDOUT1 Output Signal

3. Data Slice Level Output

**EFMI:** This pin inputs the EFM RF signal.

**DSLCO:** This pin outputs an error signal for correcting deviation in the data slice level of the EFM signal. This signal is used as a control signal to keep the data slice level of the EFM signal centered, by forming a negative-feedback loop with the EFM comparator.

DSLCI: This pin inputs the above error signal through a low-pass filter to the EFM comparator.

Figure 4 shows the EFM comparator circuit. The EFM RF signal is input through a capacitive coupling, and binarized by comparison with a slice level generated by the DSLC amplifier.

- When not in stop mode
- The DSLCO pin outputs the inverse of the binarized EFM signal. Even if the EFM signal is asymmetrical before slicing, an appropriate slice level is obtained by feeding the dc component of the sliced EFM signal back through an external low-pass RC filter.
- In stop mode
- The DSLCO pin outputs a square wave with a 50% duty cycle and the same period as the output at the MPX pin, and the slice level is kept at  $1/2 V_{DD}$ . This permits rapid optimization of the slice level when the device leaves stop mode, and prevents oscillation by cutting off the loop through the low-pass filter.
- 4. Control When a Defect is Detected

DEFCT: This pin inputs a disc defect detection signal. Both a digital PLL and an analog PLL are provided on-chip. Normally the analog PLL is used, because of its good error-rate characteristic, but when a defect detection signal is received at this pin, the chip switches over to its digital PLL for quick pull-in after the defect disappears. After pull-in, the chip automatically switches back to its analog PLL.

UCK: This pin outputs a clock signal for the microprocessor.

**UCKSL:** This pin selects the frequency of the microprocessor clock (UCK). The frequency is 16.9344 MHz when UCKSL is high, and 8.4672 MHz when UCKSL is low.

MCK: This pin outputs the master clock (33.8688 MHz).

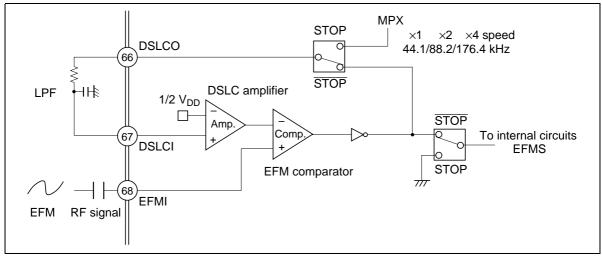


Figure 4 EFM Comparator Circuit

#### **EFM Demodulation**

After being processed in the data strobe block, the EFM signal is converted to NRZ by an NRZ-I conversion using a PLL-synchronized clock signal (PLL clock, 4.3218 MHz when the PLL is locked in standard speed playback mode).

The 24-bit frame synchronization signal is detected from this EFM signal. Operation of the EFM demodulation block is timed according to the occurrence of the frame synchronization signal.

Due to disc defects and other causes, frame synchronization signals may sometimes be detected at false positions in the EFM signal read from the disc. The sync protection block therefore opens a window around the time when the correct synchronization signal is expected, and frame synchronization signals are used for timing purposes only if they are detected within this window.

If the frame synchronization signal is not detected, it is automatically interpolated at the time when the correct frame synchronization signal would be expected to occur. Detection and interpolation of the frame synchronization signal will be described in detail in the description of the sync protection block.

After being converted to NRZ form, the EFM signal is converted to 14-bit parallel data by the EFM demodulation block. This conversion is timed to the occurrence of the above frame synchronization signal.

Next, 14-bit-to-8-bit demodulation is performed: the 14-bit parallel data is fed to the EFM demodulation ROM and converted to 8 bits.

After EFM demodulation, the 8-bit data is separated into subcode data, which is passed to the subcode signal-processing block, and audio data, which is output to the internal data bus. The data bus is connected to the error correction unit (ECU) and the RAM control block.

#### Subcode Signal Processing

**S1**: The CD format groups subcode data into 98-frame blocks. Each block begins with two subcode synchronization signals: S0 and S1. In this chip, S0 and S1 are detected in the EFM demodulation block. S0 is delayed by one frame, then ANDed with S1, and the result (S0delay·S1) is output at the S1 pin.

Due to disc defects and other causes, the above S0 and S1 signals may sometimes fail to be detected. The chip accordingly has a divide-by-98 counter that takes S0delay·S1 as its clear input and CFCKP\* as its clock input. When S0delay·S1 is not detected, it is interpolated by this counter. See figure 5.

Note: \* CFCKP is derived from the PLL clock and has a frequency of 7.35 kHz (×1 speed), 14.7 kHz (×2 speed), 29.4 kHz (×4 speed) when the PLL is in lock.

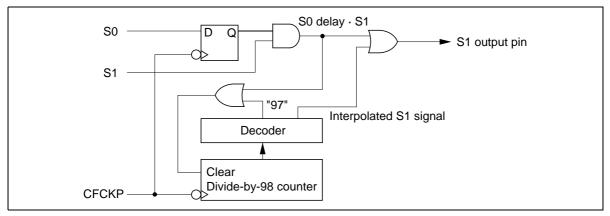


Figure 5 Block Diagram of S1 Signal Detection Circuit

**QDATA:** This is the output pin for the Q subcode data.

**QDSEL:** This pin selects one of the following two modes.

a. Q code buffer mode (selected when QDSEL is low)

When the QDSEL pin is low, the chip uses its 80-bit Q code buffer function, and outputs the Q subcode from the QDATA pin in synchronization with an external clock signal (for example, a clock signal from a microprocessor).

As shown in figure 6, the chip has two 80-bit registers. While Q code data is being written in one register, the Q code can be read from the other register asynchronously, by input of clock signals from the microprocessor at the CKEXT pin. This feature places less of a load on the microprocessor.

To switch between reading and writing of the shift registers, the S1 and QOK signals are ANDed, so before sending clock pulses for input to CKEXT, the microprocessor should check for the fall of S1, then check that QOK is high (indicating that the cyclic redundancy check of the Q data passed). These checks will enable the Q subcode to be read correctly.

The 80-bit shift register is designed to store data in 4-bit nibbles, LSB first. If the microprocessor inputs serial data in LSB-first form, it does not have to rearrange the 4 bits.

Figure 7 shows the timing chart.

b. Q code internal synchronization mode (selected when QDSEL is high)

When the QDSEL pin is high, a Q code strobe clock generated in the HD49235 is output from the CFCKP pin, and the Q code is output from the QDATA pin at a rate of one bit per frame, synchronized with the strobe clock (CFCKP). This is referred to as Q code internal synchronization mode. Figure 8 shows the timing.

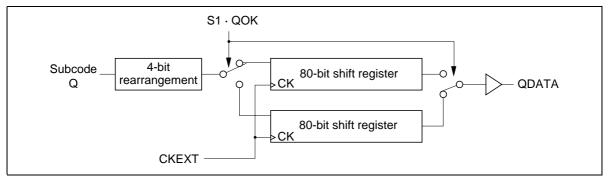


Figure 6 Block Diagram of Q Code Buffer When QDSEL is Low

QOK: The output at this pin indicates whether or not the Q subcode is correct.

One block of Q subcode data consists of 98 bits, of which 16 bits are parity bits that indicate whether the data read from the disc was correct or in error. An on-chip cyclic redundancy check circuit decides whether the 98-bit data string is correct or not, and outputs the result at the QOK pin. A high-level output indicates OK. Figures 7 and 8 show the output timing.

In modes using the Q code buffer register, when the microprocessor sets the CRCQ bit to 1, QOK is inserted in QDATA at the rise of S1.

See the note in the timing chart shown in figure 7.

**EMP:** This output pin indicates the presence or absence of pre-emphasis. The pre-emphasis signal is detected from the Q subcode and output at the EMP pin. High output indicates audio with pre-emphasis. Low output indicates audio without

pre-emphasis.

SUBOUT: This pin outputs codes R to W for use in display of graphics.

CFCKP: This pin outputs a subcode frame synchronization signal.

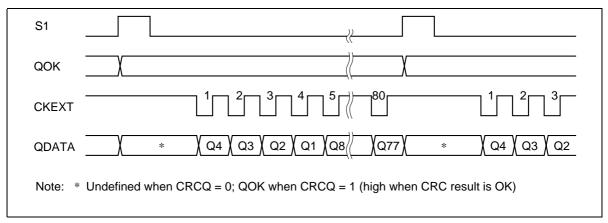
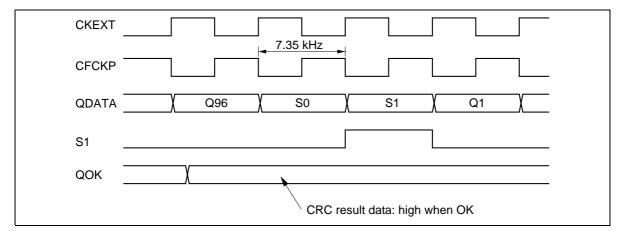
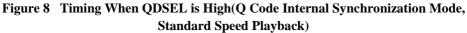


Figure 7 Timing When QDSEL is Low (Q Code Buffer Mode)





SUBCK: This pin inputs a subcode read clock.

Codes R to W are output together with codes P and Q. The codes are output in order, starting with the P code, as serial data from the SUBOUT pin when read clock pulses are input at the SUBCK pin. Figure 9 shows the timing, which basically conforms to EIAJ CP-2401.

**Signal Configuration:** Figure 9 shows the signals output for use in display of graphics. (SF: subcode frame)

To read the subcode data, eight subcode clock pulses (SUBCK) should be input after the fall of the subcode frame synchronization signal (CFCKP). The data for subcode channel P is output at the fall of CFCKP. Data in channels Q to W is output at the rise of SUBCK. See figure 10.

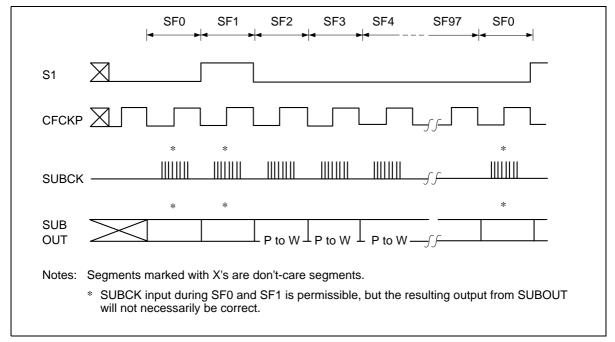


Figure 9 Output Timing for Graphics Display (1)

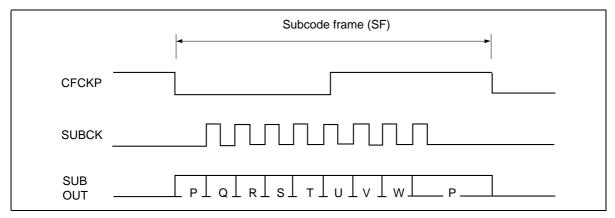


Figure 10 Output Timing for Graphics Display (2)

#### Control of 16-kbit On-Chip SRAM

The demodulated EFM data is synchronized with the PLL clock, and its output timing may contain jitter due to disturbances in the CLV servo that controls disc rotation. To absorb the jitter, the demodulated EFM data is stored in the on-chip RAM, then read out in synchronization with a clock signal derived from the crystal oscillator. The RAM capacity sets a limit on the amount of jitter that can be absorbed. In this chip, a delay of  $\pm 5$  frames between RAM read and write would lead to overwriting of existing data. The overwritten data would be destroyed, making the reproduced sound unreliable.

To avoid this, if the read and write base counters get more than  $\pm 5$  frames out of step, the write base counter is set to the value of the read base counter and the frame jitter margin is set to the maximum,  $\pm 5$  frames.

**OVFW:** This pin outputs a high RAM overflow flag signal to indicate that the difference between the read and write base counters exceeded  $\pm 5$  frames and the write base counter was set to the value of the read base counter.

MUTE: This pin is used to force the audio data to the mute state.

When MUTE is low, muting is not performed.

When MUTE is high, muting is performed.

When MUTE goes high, the address control circuit is initialized so as to maximize the RAM frame jitter margin at that point. This initialization is performed continuously while MUTE is high. Normal reproduction resumes when MUTE goes low.

#### **Error Correction Unit (ECU)**

The error correction unit can correct two-symbol C1 errors and four-symbol C2 errors.

The results of C1 error correction are flagged by a C1 flag. Since two-symbol errors can be corrected, each C1 correction produces a 2-bit C1 flag. The C1 flag data is written into an internal buffer RAM area and is read out again during C2 correction.

C2 error correction is carried out using the calculated error locations and error values, and the C1 error status and error positions indicated by the C1 error flags.

The interpolation block reads audio data and the corresponding C1 and C2 flags. If it decides from the C1 and C2 flags that the audio data is unreliable, it performs mean-value interpolation or preceding-value interpolation.

**TC1:** This pin outputs a signal indicating whether each frame of data read from the disc contained an error. See figure 11 for the output timing.

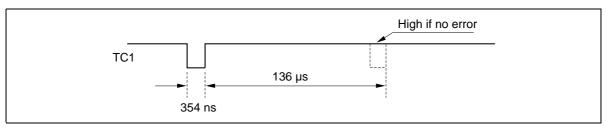


Figure 11 TC1 Timing (Standard Speed Playback)

#### **Microprocessor Interface**

DATA: Input pin for receiving microprocessor command data.

CLK: Clock input pin for receiving microprocessor command data.

**XLT:** Latch clock input pin for storing microprocessor command data in an internal register after serial input.

XRST: Input pin for clearing the microprocessor command registers.

**SENS:** This output pin provides the microprocessor with the following servo information. For details, see the microprocessor command descriptions.

- The SENS signal goes low when the number of pulses input at the CNIN pin reaches a value set by the microprocessor. Alternatively, SENS toggles between low and high each time this value is reached.
- When the constant linear velocity (CLV) servo operates in brake mode, SENS goes low to indicate detection of an interval of 32 T or more. This indicates that braking has operated and the velocity has fallen to 1/3 or less.

**CNIN:** This pin receives track-crossing pulses from the servo IC, so that the number of tracks can be counted.

1. Data Transfer Format

The microprocessor interface transfers serial data using three signal lines: XLT, CLK, and DATA. See the timing diagram in figure 12. D11 to D8 specify a register address and D7 to D0 give bit values to be set in that register.

- Notes: 1. When the external reset input signal (XRST) goes low all registers are reset to their default values. See table 1.
  - 2. Always write 0 in the following register bits:
    - D2 in register 8
    - D3 in register E
    - D7 and D6 in register F
  - 3. Always write 1 in the following register bit:
    - D7 in register 9

#### Table 1Default Values

Register Code	D7	D6	D5	D4	D3	D2	D1	D0
8 (1000)	0	0	0	0	0	0	0	0
9 (1001)	1	0	0	0	1	1	0	
A (1010)	0	0	0	0	0	0	0	0
B (1011)	0	0	0	0	0	0	0	0
C (1100)	1	1	1	0	0	0	0	1
D (1101)	0	0	0	0	0	0	0	
E (1110)	0	0	0	0	0			
F (1111)	0	0	0					

DATA	D0	D1	D2	D3	D4	D5	D6	D7	D8	D9	D10	D11	
XLT													

Figure 12 Microprocessor Interface Timing

2. Microprocessor Commands

These commands are summarized in the microprocessor command tables. Further details and notes are given below.

a. Mode Selections (Register 8)

**ROM:** This bit controls whether or not interpolation is performed on audio data output from DAS.

When ROM = 0, interpolation is performed (for audio applications).

When ROM = 1, interpolation is not performed (for CD-ROM applications).

**ROMEF:** When ROM = 1 (for CD-ROM applications), the upper and lower C2 flag data is output in two 8-bit segments. ROMEF selects which is output first: the upper or lower data.

When ROMEF = 0, the lower data is output first.

When ROMEF = 1, the upper data is output first.

DOOFF: This bit switches the digital audio interface output from pin 17 (BIDAT) on or off.

When DOOFF = 0, a signal is output.

When DOOFF = 1, the BIDAT pin is in the high-impedance state.

**SUBCO:** This bit selects whether to insert subcode data in the DAS output on the DAC output interface (MPX, CKX, DAS) in 48fs clock mode (when microprocessor command bit SLTSW is 0).

When SUBCO = 0, subcode data is not inserted.

When SUBCO = 1, subcode data is inserted.

SLTSW: This bit selects 48fs clock mode or 64fs clock mode for the DAS output.

When SLTSW = 0, DAS data is output in 48fs clock mode.

When SLTSW = 1, DAS data is output in 64fs clock mode.

**DCOND:** This bit selects the condition for switching from the analog PLL to the digital PLL when a defect is detected, in terms of the width of the defect detection signal input at the DEFCT pin (pin 69).

When DCOND = 0, the width must be at least four frames.

When DCOND = 1, the width must be at least eight frames.

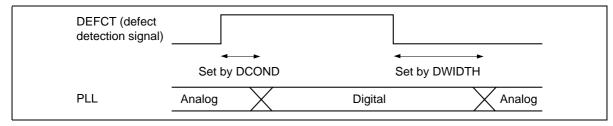


Figure 13 PLL Modes when a Defect is Detected

**DWIDTH:** This bit selects the interval from the high-to-low transition of the defect detection signal until termination of the digital PLL (and return to the analog PLL).

When DWIDTH = 0, termination occurs 8 frames past the fall of the defect detection signal if the defect detection signal width was less than 12 frames, and 16 frames past the fall of the defect detection signal if the defect detection signal width was 12 frames or more.

When DWIDTH = 1, termination occurs 4 frames past the fall of the defect detection signal if the defect detection signal width was less than 12 frames, and 8 frames past the fall of the defect detection signal if the defect detection signal width was 12 frames or more.

b. Function Selections (Register 9)

BI0 and BI1: These bits select normal play, double-speed play, or quadruple-speed play.

Note: When changing the playback speed, always switch the system to CLV stop mode so that the freerunning frequency is adjusted automatically. The period of setting is 30 msec or more.

WG10TL: This bit selects the width of the frame synchronization signal (SYNC) detection window.

WG10TL = 0: The window width is  $\pm 10$  T.

WG10TL = 1: The window width is  $\pm 19$  T.

**SYLCK0 and SYLCK1:** These bits select whether the sync protection state is maintained for 2, 4, 8, or 12 consecutive frames in which the frame synchronization signal (SYNC) is not detected.

**CRCQ:** In Q code buffer mode (when the QDSEL signal at pin 65 is low), this bit selects whether or not to insert the QOK flag into the Q code data.

c. Audio Control (Register A)

The audio control commands concerning the DAS output are all ignored in CD-ROM mode, which is selected when the ROM microprocessor command bit is set to 1.

**MUTEL and MUTER:** These command bits mute the left and right channels independently. Attenuation is carried out in eight steps (7/8, 6/8, 5/8, 4/8, 3/8, 2/8, 1/8, 0) with 136  $\mu$ s per step (at standard speed).

Notes: 1. Muting begins as soon as the command is input, without waiting for a zero-crossing point.

2. If the external MUTE signal is high, both channels are muted regardless of these commands.

MONO: This bit selects monaural audio output.

Note: When MONO is set to 1, mean-value interpolation is not performed. The only type of interpolation performed is to hold the preceding value.

ATT: This bit attenuates the audio output level by -12 dB in six steps (7/8, 6/8, 5/8, 4/8, 3/8, 2/8).

Note: If the external MUTE signal is high, both channels are muted regardless of this command bit.

**BLGMAIN and BLGSUB:** These bits select whether or not to output bilingual audio on the left and right channels. This command is ignored if "SWLR" = 1.

**SOFTMT:** This command bit mutes both the left and right channels simultaneously. Attenuation is carried out in eight steps with 136 µs per step (at standard speed).

Note: If the external MUTE signal is high, both channels are muted regardless of this command bit.

SWLR: Reverses the left- and right-channel outputs, by reversing DATA from RAM.

Note: Clear this bit to 0 when using DAS subcode output.

d. Track Counter Setting (Register B)

An internal counter counts the track-crossing signal input at the CNIN pin. When the count reaches the value set in register B, the SENS output inverts.

Depending on the order in which registers B and C are set, tracks are counted in complete mode (once only) or count mode (repeatedly). See the timing diagram in figure 14.

- Complete mode
  - Step 1: Set desired values in all registers other than registers B and C.
  - Step 2: Set register C.
  - Step 3: Set the count value in register B.
  - Step 4: Monitor the SENS line at the microprocessor.

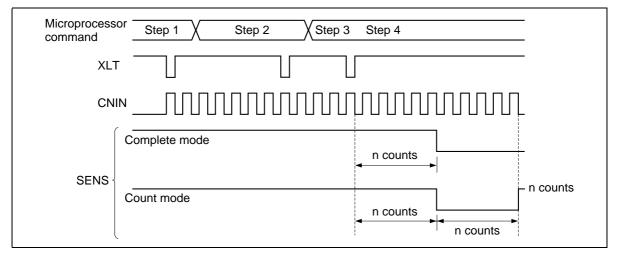


Figure 14 Track Counting

- Count mode

Step 1: Same as complete mode.

- Step 2: Set the count value in register B.
- Step 3: Set register C.
- Step 4: Monitor the SENS line at the microprocessor.
- Notes: 1. Do not use the microprocessor interface while the microprocessor is monitoring the SENS line.
  - 2. If all zeros are written in register B, the count setting is 256.
  - 3. Do not monitor the SENS pin between steps 2 and 3.

#### e. CLV Control (Register C)

**AINTV:** When a disc is played in CLV mode, the switchover between the starting servo and normal servo modes is made automatically by testing the number of sync pulses per interval. The AINTV bit selects the length of the interval.

**ATH:** When a disc is played in CLV mode, the switchover between the starting servo and normal servo modes is made automatically by testing the number of sync pulses per interval. The ATH bit selects the threshold number of pulses. Starting servo mode is used if the number of sync pulses detected in the interval selected by AINTV is less than the threshold value selected by ATH. Normal servo mode is used if the number exceeds this threshold. Only sync pulses that are validated by the sync protection function are counted.

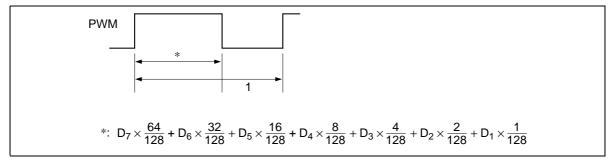
**GAIN1 and GAIN0:** These command bits select the gain of the output at the PWM pin in normal CLV servo mode. There are three selections: -6 dB, 0 dB, and +6 dB.

**SGAIN1 and SGAIN0:** These bits select the PWM gain to be one of three values, -6 dB, 0 dB or 6 dB, in start mode.

**PDGAIN1 and PDGAIN0:** These bits select the MSTOP pin output (CLV phase error) gain to be one of three values, -6 dB, 0 dB or 6 dB.

f. CLV Kick Control (Register D)

**KICK7 to KICK1:** When kick control is enabled (by microprocessor command register E) in CLV mode, these bits select the CLV control output pin PWM duty cycle to be one of 128 levels. For example, to set the duty to be 74/128, set bits D7 to D0 in microprocessor control register D to be 1001010 (base 2).



#### Figure 15 CLV Kick Control Output

#### g. CLV Mode (Register E)

This register determines the constant linear velocity control mode. Command data written in bits D7 to D4 (ED3 to ED0) of register E selects stop, play, rotate, kick, brake, access, start, or normal mode. For details of these operating modes, see the description of the CLV servo block.

h. ECU Mode (Register F)

The error-correcting capability of the error-correcting unit (ECU) can be selected. Correction when a track jump occurs can be limited to two symbols at the C1 level and three symbols at the C2 level by setting bit AS0 to 1, to reduce the likelihood of false corrections.

#### Sync Protection Block

The pulse width of the EFM signal read from the disc is measured, using the crystal oscillator clock as a time base. The pulse width value is used to detect the synchronization pattern consisting of the first 24 bits in each frame, and produce a synchronization pulse named ASYNC.

Due to disc defects and other causes, ASYNC may be detected in an incorrect position, so a divide-by-576 counter\* is used to establish a window, and only ASYNC pulses detected within this window are regarded as valid synchronization pulses (referred to as valid ASYNC pulses). Other synchronization pulses are disregarded. The width of this window can be set to one of two values by the microprocessor command WG10TL as follows.

WG10TL = 0: The window width is set to  $\pm 10T (\pm 1.7\%)$ 

WG10TL = 1: The window width is set to  $\pm 19T (\pm 3.1\%)$ 

Valid ASYNC pulses occur with correct synchronization timing, but they may sometimes be missing, e.g. because of rejection of pulses outside the detection window. Where valid ASYNC pulses are missing, PSYNC pulses are generated by interpolation. Valid ASYNC and PSYNC are the basic constant linear velocity control signals used in the CLV motor control circuit.

When two consecutive valid ASYNC pulses are detected, the chip assumes that it has acquired synchronization lock and drives the SLOCK pin high.

If valid ASYNCs are not detected for a consecutive number of times set by the SYLCK0 and SYLCK1 microprocessor command bits, interpolation is stopped and the SLOCK pin is driven low.

When (SYLCK1, SYLCK0) are (0, 0): if ASYNCs are missing for 2 consecutive times interpolation is stopped and the SLOCK pin goes low.

When (SYLCK1, SYLCK0) are (0, 1): if ASYNCs are missing for 4 consecutive times interpolation is stopped and the SLOCK pin goes low.

When (SYLCK1, SYLCK0) are (1, 0): if ASYNCs are missing for 8 consecutive times interpolation is stopped and the SLOCK pin goes low.

When (SYLCK1, SYLCK0) are (1, 1): if ASYNCs are missing for 12 consecutive times interpolation is stopped and the SLOCK pin goes low.

Figure 17 is a timing diagram for the valid ASYNC and SLOCK relationships for the case where detection of two consecutive missing valid ASYNCs was specified by microprocessor command.

Note: \* 4.2336 MHz/576 = 7.35 kHz (standard speed playback)

**SLOCK:** This output pin indicates whether sync signals were detected correctly during disc playback. This signal goes high when correct sync signals are detected in two consecutive frames, and goes low when sync signals are missing consecutively for the number of times specified by the SYLCK0 and SYLCK1 microprocessor command bits.

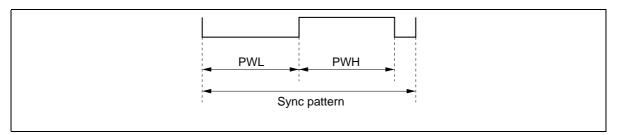


Figure 16 Sync Pattern at Start of Frame

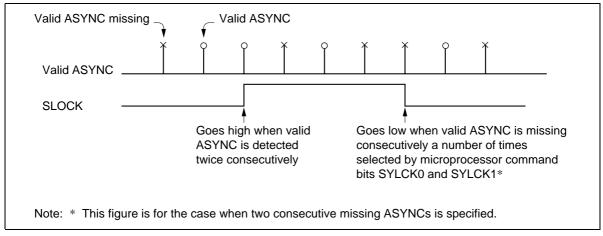


Figure 17 Valid ASYNC and SLOCK Relationships

#### **CLV Servo Control**

Compact discs (CDs) are recorded at a constant linear velocity (CLV). This block performs CLV motor control.

**MSTOP:** This CLV phase error output pin either is in the high-impedance state, or outputs a constant low signal, a constant high signal, or a pulse-width modulated waveform with a duty cycle of 0 to 100%, depending on the CLV operating mode.

**PWM:** This CLV velocity error output pin either is in the high-impedance state, outputs a constant low signal, outputs a constant high signal, or outputs a pulse-width modulated waveform with a duty cycle of 0 to 100%, depending on the CLV operating mode.

**MON:** This output signal indicates when the disc motor is on. When the disc motor is on, this pin is at the high level, except in stop mode.

PW64: This pin outputs the brake release signal.

**ROTD:** This pin can be used to monitor the most significant bit of the 7-bit (128-step) output at the PWM pin.

**CLVS:** This output pin differentiates between the starting and normal CLV modes. High output indicates normal mode.

Next the operating modes will be described. Table 2 indicates the CLV control output states in each mode.

		Outputs Signals						
CLV Mode	ED3 to 0	MON	MSTOP (Phase Error)	PWM (Speed Error)				
STOP	0000	L	Z	Z				
PLAY	0110	Н	H 50% (starting mode) 0 to 100% 0 to 100% (normal mode)					
ROT	1000	Н	50%	Н				
KICK	1001	Н	50%	Set by microprocessor				
BRAKE	1010	Н	50%	L				
ACS	1100	Н	50%	0 to 100%				
START	1110	Н	50%	0 to 100%				
NORM	1111	Н	0 to 100%	0 to 100%				

#### Table 2 CLV Control Output

#### 1. Stop Mode

This is the state in which the motor is stopped. The free-running frequency of the data strobe VCO is automatically adjusted in this mode.

#### 2. Play Mode

This mode is automatically subdivided into a starting mode (rough servo control) and normal mode (fine servo control) according to the rate at which sync signals are detected from the disc. The AINTV and ATH bits in register C (CLV control) select an interval length and threshold value. Starting mode is used if the number of sync signals in the selected interval is less than the threshold value. Normal mode is used if the number exceeds the threshold value. Starting and normal modes operate as follows.

**Starting Mode:** See table 2 for the outputs at the MON and MSTOP pins. The PWM pin outputs a rectangular wave with a duty cycle corresponding to the arithmetic mean of the pulse width on the 11-T low side and 11-T high side in the sync signal detected from the disc. See the PWM output duty cycle characteristic in figure 18.

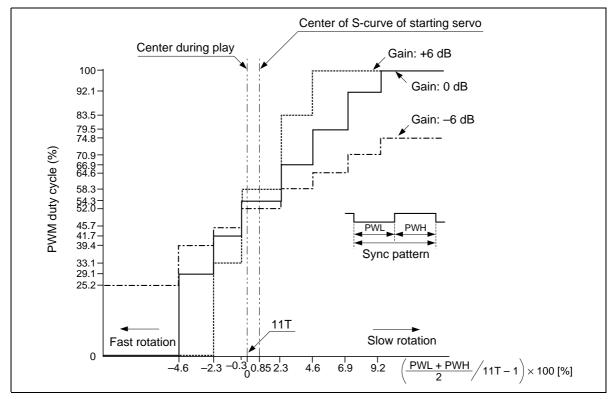


Figure 18 Duty Cycle Characteristic of PWM Output (Speed Error) in CLV Starting Mode

**Normal Mode:** The MON output pin is held high. The PWM pin outputs a rectangular wave with a duty cycle that depends on the length of four cycles of the sync signal reproduced from the disc (four frames). See the PWM output duty cycle characteristic in figure 19. The MSTOP pin outputs a phase error signal obtained by comparing the phase of a signal obtained by prescaling the sync signal by a factor of four (to give a cycle length of four frames) with the phase of an internal reference signal. See the MSTOP output duty cycle characteristic in figure 20.

3. Rotate Mode

The MON and PWM outputs are held high. The MSTOP pin outputs a square wave with a 50% duty cycle. This mode is used to force the disc motor to rotate.

4. Kick Mode

The MON output pin is held high. The MSTOP pin outputs a square wave with a 50% duty cycle. The PWM pin outputs a rectangular wave with a duty cycle that can be set to any value from 0 to 100% in 128 steps by setting bits KICK7 to KICK1 in microprocessor command register D.

5. Brake Mode

The MON output is held high. The MSTOP pin outputs a square wave with a 50% duty cycle. The PWM output is held low. This mode is used to force the disc motor to rotate in reverse. In brake mode, the interval between edges of the EFM signal is measured. If the interval is 32 T or more (T = 1/4.3218 MHz) the SENS output is driven low and the PW64 output is driven low. These outputs can be monitored to find if braking has been effective.

6. Access (ACS) Mode

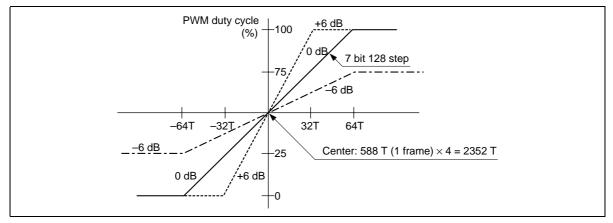
Operation is the same as in the starting submode of play mode. This mode is used in track access.

7. Start mode

Operation is the same as in the starting submode of play mode.

8. Normal Mode

Operation is the same as in the normal submode of play mode.





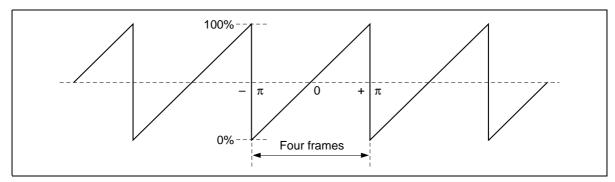


Figure 20 Duty Cycle Characteristic of MSTOP Output for Constant Linear Velocity Control

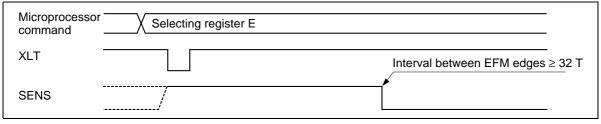


Figure 21 SENS Output in Brake Mode

#### **Digital Audio Interface**

**BIDAT:** Digital audio interface output pin. The output can be switched on or off by microprocessor command DOOFF (in register 8).

#### Interpolation

A microprocessor command SLTSW (register 8) can select the 48-fs or 64-fs clock. Figures 22 and 23 show the output timing. With the 48-fs clock, a microprocessor command SUBCO (register 8) can select whether or not to insert subcode data in the DAS output. Figures 24 and 25 show the output timing.

**DAS:** This pin outputs audio or ROM data. The ROM microprocessor command (in register 8) can select whether or not to carry out interpolation. Preceding-value and mean-value interpolation are carried out. With a 48-fs clock, the data is output MSB first and squeezed to the rear. With a 64-fs clock, the data is output LSB first and squeezed to the rear.

CKX: This pin outputs the data transfer clock.

**MPX:** This pin outputs a signal that distinguishes between the left and right channels. With a 64-fs clock, low output indicates the left channel and high output indicates the right channel. With a

48-fs clock, this polarity is reversed.

QMX: This pin outputs a clock signal with four times the frequency of MPX.

DMX: This pin outputs a clock signal with two times the frequency of MPX.

**C2F:** This output pin goes high to flag data errors that could not be corrected by C2 error correction. C2F is low when there are no such uncorrectable errors. C2F is synchronized to the data output on the DAS line. When audio is selected by the ROM microprocessor command, this signal is output once for every two bytes on the left and right channels. When CD-ROM is selected, the signal is output once per byte (every 8 bits).

See figure 26 for the output timing.

MUTE: Selects whether to mute the audio data.

Low: Muting is not performed.

High: Muting is performed.

When MUTE goes high, the address control circuit is initialized so as to maximize the RAM frame jitter margin at that point. This initialization is performed continuously while MUTE is high. Normal reproduction resumes when MUTE goes low.

#### **Other Pins**

**MRST:** Master reset. The HD49235FS chip resets when MRST goes low, and operates normally when MRST is high. This pin has a pull-up resistor, so it can be either left open or connected to  $V_{DD}$ .

**TEST1 to TEST3:** These pins input test control signals. These pins have pull-up resistors, so they can be either left open or connected to  $V_{DD}$ .

**V**<sub>DD</sub>: Power supply pin.

V<sub>ss</sub>: Ground pin.

NC: These pins should be left unconnected. Correct operation is not assured if they are connected.

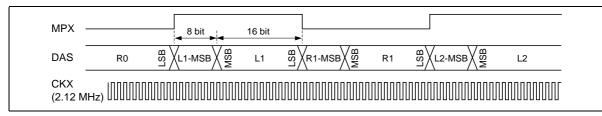


Figure 22 Audio/ROM Data Output Sequence (When 48-fs Clock is Selected): DAS Switched at Fall of CKX

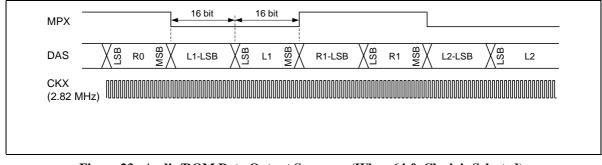


Figure 23 Audio/ROM Data Output Sequence (When 64-fs Clock is Selected): DAS Switched at Rise of CKX

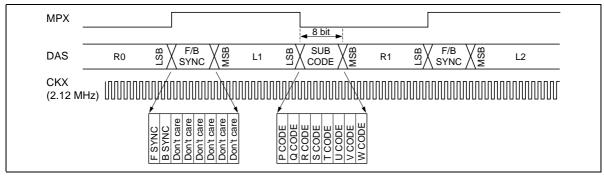


Figure 24 Audio/ROM Data Output Sequence with Subcode Data Inserted (When 48-fs Clock is Selected): DAS Switched at Fall of CKX

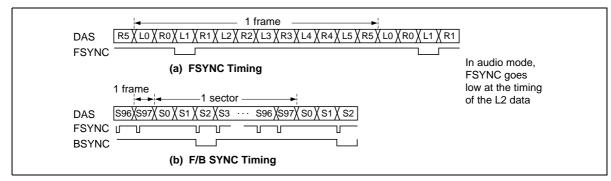


Figure 25 F/B SYNC Signal Timing Diagram

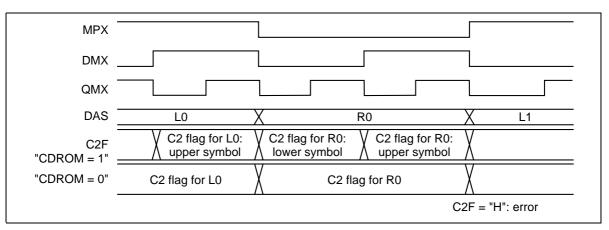


Figure 26 Error Flag Output (Switchable by Upper/Lower-1st Command in ROM Mode)

## Absolute Maximum Ratings (Ta = $25^{\circ}$ C, V<sub>ss</sub> = 0 V)

Item	Symbol	Value	Unit
Supply voltage	V <sub>DD</sub>	-0.3 to +7.0	V
Pin voltage*	V <sub>T</sub>	-0.3 to V <sub>DD</sub> + 0.3	V
Allowable power dissipation	Pd	450	mW
Operating temperature	Topr	0 to +75	°C
Storage temperature	Tstg	-55 to +125	°C

Note: \*  $V_{DD}$  + 0.3 V < 7.0 V

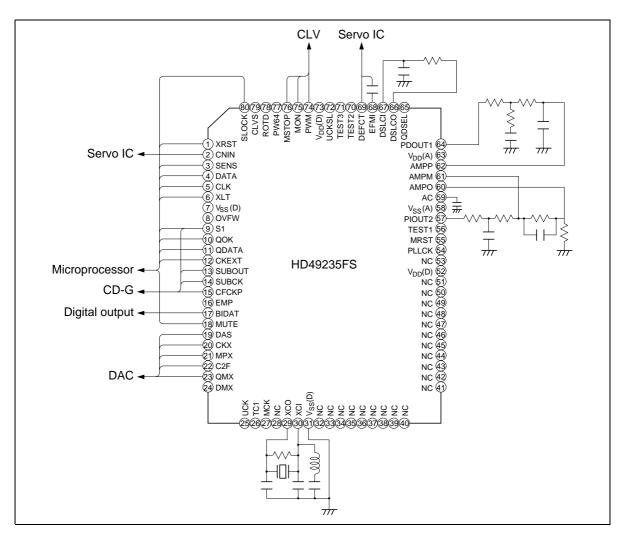
### **Electrical Characteristics**

ltem	Symbol	Min	Тур	Max	Unit	Test Conditions	Applicable Pin(s)
Operating supply voltage	V <sub>dd</sub>	4.5	5.0	5.5	V	Normal-speed, double- speed, or quadruple-speed play	52, 63, 73
Input voltage (high)	V <sub>IH</sub>	$0.7  imes V_{\text{DD}}$	_	_	V		*1
Input voltage (low)	V <sub>IL</sub>	_	_	$0.3  imes V_{_{DD}}$	V		-
voltage 1	V <sub>OH1</sub>	V <sub>DD</sub> -0.5	_	_	V	-Ι <sub>οн</sub> = 0.6 mA	*2
Output low voltage 1	$V_{\rm ol1}$	_	_	0.4	V	I <sub>oL</sub> = 0.6 mA	-
Input leakage current	I <sub>IL</sub>	-5	0	5	μA		*1
Three-state leakage current	I <sub>TOL</sub>	-5	0	5	μA	Three-state output pins in high-impedance state	3, 17, 74, 75
Input pin pull-up resistance	Rip	10	20	40	kΩ		55, 56, 65, 72
Amplifier output	V <sub>AO</sub>	2.4	2.5	2.6	V	V <sub>AP</sub> input (at AP pin): 2.5 V	60
VCO output	PLLCK	4.2	4.7	5.2	MHz	Normal speed	54

Notes: 1. Input pins and I/O pins in input mode (except analog pins): 1, 2, 4, 5, 6, 12, 14, 18, 55, 65, 67, 69, 72.

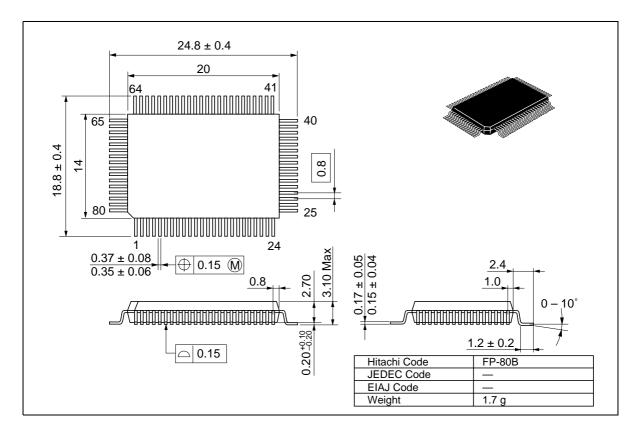
2. The following output pins, I/O pins in output mode, and three-state output pins: 3, 8 to 11, 13, 15, 16, 17, 19 to 27, 57, 64, 66, 74 to 80

## **Application External**



## **Package Dimensions**

Unit: mm



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